

(D) The Parry "Cathamplifier"

The basic circuit is Fig. 13.46A, the two cathodes being coupled by a centre-tapped a-f transformer, whose secondary winding excites the grid of V_2 . A theoretical analysis is given in Ref. E29, while some practical designs are in Ref. E30.

$$\text{For balance, } \frac{N_3}{N_1 + N_2} = \frac{1 + g_m R / 2}{g_m R} \text{ and } N_1 = N_2$$

where g_m = mutual conductance of V_2 .

Distortion is reduced by the factor $T(2T - 1)$

where $T = N_3 / (N_1 + N_2)$.

Note that T should normally be slightly greater than 1.

Gain is reduced by the factor $T(T - 0.5)$.

The common cathode resistor R_0 helps to reduce unbalance.

In practice, R is made variable (say 100 ohms total) so as to permit the amplifier to be balanced experimentally. One method is to connect a valve voltmeter across R_0 , and to adjust R for minimum reading.

Instability may occur if R is too small.

A modified circuit is Fig. 13.46B in which the centre-tapped primary of T_1 is not necessary.

Fig. 13.46C permits both a.c. and d.c. balancing.

Fig. 13.46D keeps the circulating screen current out of the cathode circuit and so maintains the ratio between plate and screen currents at the negative voltage peak swing. Resistors R_1 are to prevent coupling from cathode to cathode through the screen by-pass condensers; their values should be low—say 100 to 250 ohms each.

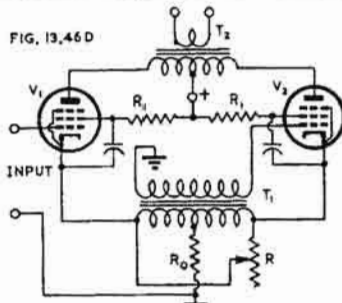
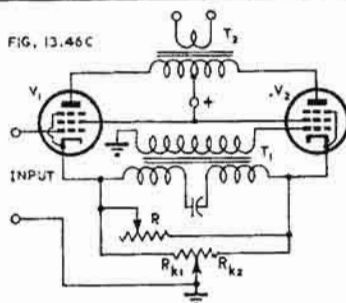
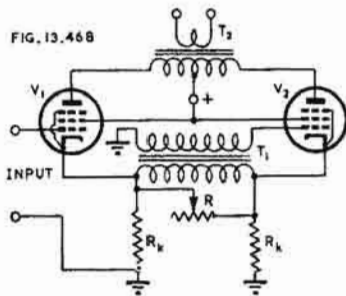
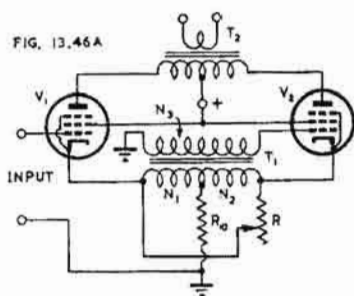


Fig. 13.46A. Basic circuit of Parry Cathamplifier, (B) Modified circuit, (C) With both a.c. and d.c. balancing, (D) Keeps circulating screen current out of cathode circuit (Ref. E30).